

#### Lecture 8

#### Spread Spectrum and OFDM



### Time Domain View (Sieve)



Direct Sequence Spread Spectrum

# Spread Spectrum

- Usually the spectrum of a signal is related to the data (symbol) rate
  - The null-to-null bandwidth  $\cong 1/T$
  - T is the symbol duration

#### Spread-spectrum

- The spectrum is much wider than 1/T
- The spreading is achieved using a "spreading signal" also called a "code signal" or "spreading code"
- The receiver uses correlation or matched filtering to recover the original data

### Types of Spread Spectrum

#### Direct-sequence spread spectrum (DSSS)

Sach information symbol is "chipped" into a pattern of smaller

Rtern is called the spread-spectrum "code" or "sequence"

sed in IS-95, W-CDMA, cdma2000 and IEEE 802.11

#### Frequency hopping spread spectrum (FHSS)

- Symbols or packets are transmitted on different frequency carriers each time
- Slow frequency hopping the same frequency carrier is used over several symbols or a packet (common)
- Fast frequency hopping the frequency carrier is changed within a symbol period
- Used in GSM, IEEE 802.11 (legacy) and Bluetooth

### Systems using Spread Spectrum

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DSSS is employed in 2G CDMA systems

- IS-95, cdma2000
- DSSS is employed in all 3G cellular systems
   UMTS and HSPA
- DSSS was used in legacy IEEE 802.11 (WiFi)



- The original data stream is "chipped" up into a pattern of pulses of smaller duration
- Good autocorrelation properties
- Good cross-correlation properties with other patterns
- Each pattern is called a spread spectrum code or spread spectrum sequence



Periodic Spreading Code

### DSSS details

- Instead of transmitting a rectangular pulse for a zero or a one, we transmit a sequence of narrower rectangular pulses
- The narrow pulses are called "chips"
  - You often see references to "chips/sec" instead of bits/sec
- The easiest way of creating a DSSS signal is to multiply one period of the spreading sequence with each data symbol
  - Example: IEEE 802.11
    - Barker sequence: [1 1 1 -1 -1 1 -1 -1 1 -1 ]
    - To transmit a "0", you send [1 1 1 -1 -1 -1 1 -1 -1 1 -1]
    - To transmit a "1" you send [-1 -1 -1 1 1 1 -1 1 1 -1 1]
  - Sometimes parts of the spreading sequence are multiplied with the data symbol

# Processing gain

- Definition of processing gain
  - The duration of a chip is usually represented by  $T_c$
  - The duration of the bit is T
  - The ratio  $T/T_c = N$  is called the "processing gain" of the DSSS system
- The processing gain is also the ratio between the bandwidth of the spread signal to the bandwidth of the data signal
- In many cases, this is also the ratio of the height of the autocorrelation peak to the maximum sidelobe
  - This ratio depends on the spreading code properties

# Operation of a DSSS Transceiver



Demodulation involves a process called "correlation"

### Spectrum and Autocorrelation



#### Autocorrelation properties of the Barker sequence

The width of the mainlobe is 2*T*/11

- About one-tenth the width of the autocorrelation of the rectangular pulse
- The height of the mainlobe is 11 times the height of the sidelobes
- The ratio of mainlobe peak to sidelobe is an important measure of how "good" a spreading code is

# + 7- Chip M-sequence



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### + Example in a two-path channel

- Random data sequence of ten data bits
  - Spreading by 11 chips using a Barker pulse
- Two path channel with inter-path delay of 17 chips > bit duration
- Multipath amplitudes
  - Main path: 1
  - Second path: 1.1
- Just for illustration!
- Reality:
  - Many multipath components
  - Rayleigh fading amplitudes
  - Noise!







#### Without Multipath

With Multipath



### Summary of DSSS and Combatting Multipath



## The RAKE receiver

- Observe the peaks in the channel output in the previous slides that are NOT sampled (Peaks that are not at the green vertical line)
  - They contain the "same" information as the sampled peaks but these peaks are delayed!
- A RAKE receiver consists of a tapped delay-line that samples these peaks
- Each peak usually suffers independent fading
  - This is a form of diversity inherently available in DSSS systems
- In IS-95 systems the RAKE receiver has three "fingers"
  - It can sample three such peaks simultaneously
  - A 4th finger is used to listen to adjacent cells for RSS measurements and to support soft hand-off
    - The mobile station is temporarily connected to more than one base station

### Principle of RAKE Receiver

#### Steps

- Multiple versions of a signal arrive more than one chip interval apart
- Receiver attempts to recover signals from multiple paths and combine them
- This method achieves better performance than simply recovering dominant signal and treating remaining signals as noise



# CDMA/DSSS Summary



#### CDMA Properties: Near-Far Problem

- A CDMA receiver cannot successfully de-spread the desired signal in a high multiple-access-interference environment
- Unless a transmitter close to the receiver transmits at power lower than a transmitter farther away, the far transmitter cannot be heard
- Power control must be used to mitigate the near-far problem
- Mobiles transmit at such power levels to ensure that received power levels are equal at base station

Power control and channel problems!



### CDMA Deployment Issues

Radio planning in CDMA systems is different from standard TDMA/FDMA systems

- Reuse is defined differently
- Capacity calculations are different

### Network planning for CDMA

- There is no concept of co-channel or adjacent channel interference
  - Interference arises from users in the same cell and from neighboring cells
  - Coding and spread spectrum play a very important role in the mitigation of interference
- Instead of defining an  $S_r$  based on signal strength, it is more common to use a value of  $E_b/I_t$  that provides a given "quality of signal"
  - Usually this is the value that provides a frame error rate of 1% – this provides a good MOS for voice
  - The quantity I<sub>t</sub> is the total interference

## More on $E_b/I_t$

#### • The value of $E_b/I_t$ depends greatly on

- Propagation conditions
- Transmit powers of the interfering users
- Speed of the MS
- Number of multipath signals that can be used for diversity

#### Cell breathing

- The boundary of a CDMA cell is not fixed and depends on where the  $E_b/I_t$  is reached
- Capacity must be offloaded to other carriers to overcome this effect



- Power control, soft handoff and RSS thresholds play a very important role in the design
  - If too many BSs (or sectors) cover an area, this may create a "coverage hole"
  - Usually, not more than three BSs or sectors should cover an area

### Approach

Somewhat simplified, but works in general for *M* users in a cell

- Let us consider the reverse link (uplink)
  - There are two components of the interference
    - Own cell interference *I*<sub>o</sub>
    - Other cell interference *I*<sub>oc</sub>
- Assuming perfect power control, the own cell interference is given by:

 $I_o = (M-1) S v_f$ 

S is the average power received from each of the M mobile stations

The reverse link "activity factor" is v<sub>f</sub>

The activity factor is a measure of what fraction of time a transmission occurs

### + Other Cell Interference

- Interference from other cells fluctuates as a function of the load
- The average value I<sub>oc</sub> can be expressed as follows

 $I_{oc} = f M S v_f$ 

- Assumption is that all other cells are similar to the current cell
- The factor f indicates fraction of other cell received power compared to the own-cell received power
- In some ways, *f* is a measure of the reuse factor
- The factor f depends on the size of the given cell, the path loss exponent, shadow fading distribution, soft handoff parameters, etc.

### Approach (II)

Total interference is given by:

 $I_{total} = I_o + I_{oc} = [(1+f)M-1] v_f S = [M/\eta - 1] v_f S$ 

- Here the term  $\eta$  refers to the "reuse efficiency"
- Suppose there is imperfect power control, we can represent this by a factor  $\eta_c$

 $I_{total} = [M/\eta - 1] v_f(S/\eta_c)$ 

In general, the required SIR must be smaller than the observed SIR

 $(E_b/I_t)_{req} < (SIR)_{system}$ 

Ignore thermal noise

The desired signal has a power S multiplied by the "processing gain" G<sub>p</sub>

Approach (III)

Proceeding further, we get:

$$\frac{E_b}{I_t} = \frac{S G_p}{[M/\eta - 1] v_f(S/\eta_c)} = \frac{S G_p}{[(1+f)M-1] v_f S/\eta_c}$$

$$M = \frac{1}{1+f} + \frac{G_{p} \eta_{c}}{(E_{b}/I_{t}) v_{f} (1+f)}$$

Solving for M we get:  

$$M_{max} = 1 + \frac{G_p \eta_c}{(E_b/I_t) v_f (1+f)}$$

*M<sub>max</sub>* is called the "pole point" or asymptotic cell capacity

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#### Cell Loading and Pole Point in IS-95

#### Cell loading

- A measure of the total interference in the system compared to thermal noise
- Represented by the quantity  $\rho = M/M_{max}$
- You can show that it is also approximately equal to the ratio of the total interference to the thermal noise

#### Sample calculation

- Let  $(E_b/I_t)_{reqd} = 6 dB = 4$ , R = 9.6 kbps,  $R_c = 1.2288 Mcps$ ,  $\eta_c = 0.8$ ,  $v_f = 0.5$ , f = 0.67
- Then, the pole point or  $M_{max}$  will be:
- $M_{max} = 1 + (1.2288 \times 10^{6}/9.6 \times 10^{3})(0.8/(4 \times 0.5 \times [1+0.67]) = 1 + 30.65 = 32$
- If a 3 sector antenna is used, typically, the gain in capacity is by a factor of 2.55 so that the pole point is: 31.65× 2.55 = 81

### Comparison with AMPS/TDMA

In AMPS, each service provider has 12.5 MHz BW

- With a 3 sector antenna, we can have a frequency reuse of 7
- There are 30 kHz channels per voice call
- Number of channels/cell =

 $(12.5 \times 10^{6} / 30 \times 10^{3}) \times (1/7) = 57$ 

- In the case of IS-136, with a 3 sector antenna, we can have a frequency reuse of 4
  - Each 30 kHz channel can carry 3 voice calls
  - Number of channels/cell =

 $(12.5 \times 10^6 / 30 \times 10^3) \times (1/4) \times 3 = 312.5$ 

- What was the pole point of IS-95?
  - 81 per carrier per cell sector
  - With 8 cdma carriers in a 12.5 MHz bandwidth, we can have up to 648 channels per cell sector
  - With 10 cdma carriers in a 12.5 MHz bandwidth, we can have up to 810 channels per cell sector

### Remarks

#### Ranges of values

- Power control inefficiency  $\eta_c$  varies between 0.7 and 0.85
- Voice activity factor v<sub>f</sub> varies between 0.4 and 0.6
- The other cell interference f varies between 0.56 and 1.28 for a path loss exponent of 4 and a standard deviation of shadow fading of 6 to 10 dB

### • Other issues

#### Forward Link

- We have to be worried about the pilot, sync, paging and traffic channels in IS-95 and many more in cdma2000 and UMTS
- The strength of the pilot channel effectively determines the size of the cell
- Interference is from clusters of high power transmitters rather than many distributed low power transmitters
- Design should try to make the forward and reverse link capacities as close to one another as possible
  - This will reduce the amount of unnecessary interference and enable smooth handoffs between cells
- PN Sequence Reuse
  - How closely should the same pilot offsets be used? (later when we do IS-95)
- How does the link budget affect the capacity?
- How does soft handoff affect the capacity?

# Frequency Domain View (Gate)



**Orthogonal Frequency Division Multiplexing** 

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### Diversity (continued) – Frequency Hopping

#### Traditional

Transmitter/receiver pair communicate on a fixed frequency channel.

#### Frequency Hopping Idea

- Noise, fading and interference change with frequency band in time
  - Move from band to band
- Time spent on a single frequency is termed the "dwell time"
- Originally developed for military communications
  - Spend a short amount of time in one frequency band
  - Prevent interception or jamming

#### Frequency Hopping Spread Spectrum





Developed during WWII by actress Hedy Lammar and classical composer George Antheil

Patent given to government

### Frequency Hopping Spread Spectrum

- Two types of systems
  - Slow Hopping
    - Dwell time long enough to transmit several bits in a row (timeslot)
  - Fast Hopping
    - Dwell time on the order of a bit or fraction of a bit (primarily for military systems)
- Transmitter and receiver must know hopping pattern or algorithm that determines the pattern before communications.
  - Cyclic pattern best for low number of frequencies and combating small-scale fading :
    - Example with four frequencies: f4, f2, f1, f3, f4, f2, f1, f3, ....
  - Random pattern best for large number of frequencies, combating co-channel interference, and interference averaging
    - Example with six frequencies: f1, f3, f2, f1, f6, f5, f4, f2, f6, ...
    - Use random number generator with same seed at both ends



# + Frequency Hopping concept



One Period of Sequence = 1 0 0 1 0 1 1

CLK	с	в	Α	f <sub>c</sub>
t <sub>o</sub>	1	0	0	f <sub>4</sub>
t <sub>1</sub>	0	1	0	f <sub>2</sub>
t <sub>2</sub>	1	0	1	f <sub>5</sub>
t <sub>3</sub>	1	1	0	f <sub>6</sub>
t <sub>4</sub>	1	1	1	f <sub>7</sub>
t <sub>5</sub>	0	1	1	f <sub>3</sub>
t <sub>6</sub>	0	0	1	f <sub>1</sub>
t <sub>7</sub>	1	0	0	f <sub>4</sub>



time

## Combatting Time Dispersion



# + Example Systems



GSM (2G Cellular)Very slow hopping

Original IEEE 802.11
 Slow hopping

#### Bluetooth

Also slow hopping frequency over 79 frequencies each 1 MHz wide

Per packet hopping

# How do you utilize the entire bandwidth?



# Orthogonal Frequency Division Multiplexing

- Idea in frequency domain:
  - Coherence bandwidth limits the maximum data rate of the channel
  - Send data in several parallel sub-channels each at a lower data rate and different carrier frequency
- Idea in time domain:
  - By using several sub-channels and reducing the data rate on each channel, the symbol duration in each channel is increased
  - If the symbol duration in each channel is larger than the multipath delay spread, we have few errors
- OFDM enables
  - Spacing carriers (sub-channels) as closely as possible
  - Implementing the system completely in digital eliminating analog VCOs



Modulation/Multiplexing technique

- Usual transmission
  - Transmits single high-rate data stream over a single carrier
- With OFDM
  - Multiple parallel low-rate data streams
  - Low-rate data streams transmitted on orthogonal subcarriers
  - Allows spectral overlap of sub-channels



- It is NOT a new technology but has found new importance because of applications
  - DSL modems where the channel is not uniform
  - Digital audio and video broadcast
  - Wireless LAN applications
    - IEEE 802.11a and HIPERLAN-2
- Fast implementation using FFT's is now possible
- Can be adaptive (used in 802.11a)
- Problems
  - Synchronization between carriers
  - Peak-to-average power (PAP) ratios
    - Requires linear amplifiers

## OFDM Advantages

Bandwidth efficiency

Reduction of ISI

Needs simpler equalizers

Robust to narrowband interference and frequency selective fading

Possibility of improving channel capacity using adaptive bit loading over multiple channels 46

### + OFDM in frequency and time domains

- Note orthogonality in both domains
- What is one "OFDM symbol"?



## OFDM Signal/Symbol



## + OFDM Symbol

- One OFDM "symbol" lasts for say T<sub>s</sub> seconds
  - The symbol consists of the sum of the individual symbols from the many sub-carriers
  - Example: Consider QPSK on each carrier
- In general
  - For N subchannels, the N samples of the *i*-th transmitted OFDM symbol can be written as

### Guard Time and Cyclic Prefix

- Guard time eliminates ISI if larger than expected delay spread occurs
- If the guard time has no signal, intercarrier interference (ICI) may occur
  - ICI is like a cross talk between subcarriers
- A cyclic prefix eliminates ICI
  - Ensures that delayed replicas of OFDM symbols always have integer number of cycles within the FFT interval
  - Maintains orthogonality between subcarriers
  - Cyclic prefix is removed at the receiver

### OFDM Transmission – basic system

N consecutive complex symbols are converted into a group of N parallel data streams, which then are modulated over orthogonal subcarriers





# Channel Partitioning for Multicarrier Modulation

- As the channel is frequency selective, it makes sense to split the channel into several smaller parts
  - Each smaller chunk is now an AWGN channel
  - Each AWGN channel provides a *different* SNR
- Question: How do we allocate transmit powers/modulation schemes to each chunk? What is the most optimal?



Water-filling algorithm Allocate more energy where the SNR is better!

+ Adaptive OFDM

- Improve channel capacity further
  - Change modulation scheme
  - Allocating bits/power per subcarrier according to the quality of each subchannel



## Adaptive Modulation

Set of Modulation Schemes  $\prec$ 

No transmission (0 bits) BPSK (1 bit/symbol) QPSK (2 bits/symbol) 8-QAM (3 bits/symbol) 16-QAM (4 bits/symbol)



#### Adaptive Modulation on Parallel Channels





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#### OFDM Based Wireless LANs – IEEE 802.11a

- Operates in the U-NII Band
  - **5.15–5.25**, **5.25–5.35**, and **5.725–5.825** GHz
- Provides multiple transmission modes/rates depending on channel conditions.
  - 6, 9, 12, 18, 24, 36, 48, and 54 Mbps
- 4 digital modulations: BPSK, QPSK, 16-QAM, 64 QAM
- Radio spectrum is divided into 8 separate segments/channels, 20 MHz each
- 52 carriers (subchannels) per channel
  - Each subcarrier has bandwidth of ~300 kHz
  - 48 for data modulation, and 4 for pilot signal

### + Recent Trends



MIMO with OFDM

- IEEE 802.11n, 802.11ac
- Data rates greater than 100 Mbps
- OFDM for wide area data services
  - LTE and WiMax
- Other PHY technologies
  - UWB with OFDM
  - MC-CDMA

### Revisiting "Data Rates" in Wireless

- Home A/V networks are expected to need 1-10 Gbps
  - Assuming a spectral efficiency of 1 bps/Hz, we need at least 1 GHz of spectrum
    - Have ignored the effects of multipath fading
  - Brute force approach
    - May not meet the technology, regulatory and cost requirements
- Can we increase the bps/Hz in wireless systems?



- Phil Edholm
  - Nortel's CTO
- Three Telecom
   Categories
  - Wireline
  - Nomadic (Portable) #
  - Wireless (Mobile)
- Data rates increase exponentially
  - There is a predictable time lag between wireless and wireline systems



### How can we increase data rates?

#### Traditional ways

- Reduce the symbol duration
  - Needs larger bandwidth
  - Leads to a wideband channel and frequency selectivity irreducible error rates
- Increase the number of bits/symbol
  - Error rates increase with M for the same  $E_b/N_0$

#### MIMO systems

- There is no need to increase the bandwidth or power
  - But what are the limitations?
- Use multiple transmit (Tx) and receive (Rx) antennas
- Increases spectral efficiency to several tens of bps/Hz

+What is MIMO?

- So far we have considered Single Input Single Output or SISO systems
  - Both transmitter and receiver have one antenna each
  - Simplest form of transceiver architecture
- Single input multiple-output (SIMO) systems
  - Receiver has multiple antennas
- Multiple input multiple output (MIMO) systems
  - Both transmitter and receiver have multiple antennas
  - Strictly: Each antenna has its own RF chain (modulator, encoder and so on)



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# Performance enhancements due to MIMO

#### Diversity gain

Ability to receive multiple copies of the signal with independent fading

#### Spatial multiplexing gain

Send different information bits over different antennas and recover the information

#### Interference reduction

Reduce the region of interference thereby increasing capacity